

On the “SIR”s (“Signal”-to-“Interference”-Ratio) in Discrete-Time Autonomous Linear Networks

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Abstract

In this letter, we improve the results in [5] by relaxing the symmetry assumption and also taking the noise term into account. The author examines two discrete-time autonomous linear systems whose motivation comes from a neural network point of view in [5]. Here, we examine the following discrete-time autonomous linear system: $\mathbf{x}(k+1) = \mathbf{A}\mathbf{x}(k) + \mathbf{b}$ where \mathbf{A} is any real square matrix with linearly independent eigenvectors whose largest eigenvalue is real and its norm is larger than 1, and vector \mathbf{b} is constant. Using the same “SIR” (“Signal”-to-“Interference”-Ratio) concept as in [4] and [5], we show that the ultimate “SIR” is equal to $\frac{a_{ii}}{\lambda_{max} - a_{ii}}$, $i = 1, 2, \dots, N$, where N is the number of states, a_{ii} is the diagonal elements of matrix \mathbf{A} , and λ_{max} is the (single or multiple) eigenvalue with maximum norm.

Index Terms

Autonomous Discrete-Time Linear Systems, Signal to Interference Ratio (SIR).

I. INTRODUCTION

This letter improves so-called “SIR” results in [5] where the author analyzes two discrete-time autonomous linear systems whose motivation comes from a neural network point of view. In [3]

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a dynamic-system-version of SIR (“Signal”-to-“Interference” Ratio) concept is introduced. Let’s first present the definition of the SIR in [4].

The SIR is an important entity in communications engineering which indicates the quality of a link between a transmitter and a receiver in a multi transmitter-receiver environment (see e.g. [1], among many others). For example, let N represent the number of transmitters and receivers using the same channel. Then the received SIR at receiver i is given by (see e.g. [1])

$$SIR_i(k) = \gamma_i(k) = \frac{g_{ii}p_i(k)}{\nu_i + \sum_{j=1, j \neq i}^N g_{ij}p_j(k)}, \quad i = 1, \dots, N \quad (1)$$

where $p_i(k)$ is the transmission power of transmitter i at time step k , g_{ij} is the link gain from transmitter j to receiver i (e.g. in case of cellular radio systems, g_{ij} involves path loss, shadowing, etc) and ν_i represents the receiver noise at receiver i . Typically, in cellular radio systems, every transmitter tries to optimize its power $p_i(k)$ such that the received SIR(k) (i.e., $\gamma_i(k)$) in eq.(1) is kept at a target SIR value, γ_i^{tgt} .

The author defines the following dynamic-system-version of the “Signal-to-Interference-Ratio (SIR)”, denoted by $\theta_i(k)$, by rewriting the eq.(1) with neural networks terminology in [4] and [5]:

$$\theta_i(k) = \frac{a_{ii}x_i(k)}{b_i + \sum_{j=1, j \neq i}^N a_{ij}x_j(k)}, \quad i = 1, \dots, N \quad (2)$$

where $\theta_i(k)$ is the defined fictitious “SIR” at time step k , $x_i(k)$ is the state of the i ’th neuron, a_{ii} is the feedback coefficient from its state to its input layer, a_{ij} is the weight from the state of the j ’th neuron to the input of the i ’th neuron, and b_i is constant.

In this paper, we improve the results in [5] where the author examines the ultimate SIR in the following two discrete-time autonomous linear systems whose motivation comes from a neural network point of view.

1) The discretized autonomous linear system:

$$\mathbf{x}(k+1) = (\mathbf{I} + \alpha(-r\mathbf{I} + \mathbf{W}))\mathbf{x}(k) \quad (3)$$

which is obtained by discretizing the autonomous continuous-time linear system in [4] using Euler method; where \mathbf{I} is the identity matrix, r is a positive real number, and $\alpha > 0$ is the step size.

2) A more general linear system with symmetric weight matrix

$$\mathbf{x}(k+1) = (-\rho\mathbf{I} + \mathbf{W})\mathbf{x}(k) \quad (4)$$

where \mathbf{I} is the identity matrix, ρ is a positive real number, and $(-\rho\mathbf{I} + \mathbf{W})$ is the system matrix.

In [5], the above linear systems with symmetric matrices and with zero noise term are examined. In this paper, we relax these two conditions, and examine the SIR in the general discrete-time autonomous linear system in eq.(5) where the weight matrix $\mathbf{A} \in R^{N \times N}$ is any real matrix whose largest (single or multiple) eigenvalue is real and its norm is larger than 1.

It's well-known that in the system of eq.(5), the eigenvalues of the weight (system) matrix solely determine the stability of the system. If the spectral radius of the matrix is larger than 1, then the system is unstable. The spectral radius of an $N \times N$ matrix is equal to $\max\{|\lambda_i|\}_{i=1}^N$ where λ_i is the eigenvalues of the matrix. In this paper, we examine the linear systems with system matrices whose spectral radius is larger than 1.

The paper is organized as follows: The results for system eq.(5) are presented in section II. Section III presents some simple simulation results, followed by the conclusions in Section IV.

II. "SIR" IN DISCRETE-TIME AUTONOMOUS LINEAR NETWORKS

In this paper, we examine the SIR in the following autonomous linear system

$$\mathbf{x}(k+1) = \mathbf{A}\mathbf{x}(k) + \mathbf{b} \quad (5)$$

where $\mathbf{x}(k) \in R^{N \times 1}$ is state vector at step k , $\mathbf{A} \in R^{N \times N}$ is any real square matrix with linearly independent eigenvectors whose largest eigenvalue is real and its norm is larger than 1, and vector \mathbf{b} is constant. In what follows, we present the main result of this paper:

Proposition:

In the discrete-time linear system of eq.(5), if the matrix \mathbf{A} has linearly independent eigenvectors, and if the (single or multiple) eigenvalue with the greatest norm is real and its norm is greater than 1, then the defined "SIR" ($\theta_i(k)$) in eq.(2) for any state i converges to the following constant within a finite step number denoted as k_T for almost any initial vector $\mathbf{x}(0)$.

$$\theta_i(k \geq k_T) = \frac{a_{ii}}{\lambda_{max} - a_{ii}}, \quad i = 1, 2, \dots, N \quad (6)$$

where λ_{max} is the (single or multiple) eigenvalue with maximum norm and a_{ii} is the diagonal elements of \mathbf{A} .

Proof:

From eq.(5),

$$\mathbf{x}(k) = \mathbf{A}^k \mathbf{x}(0) + \sum_{j=0}^{k-1} \mathbf{A}^j \mathbf{b} \quad (7)$$

where $\mathbf{x}(0)$ shows the initial state vector at step zero. It's well known that any real square matrix \mathbf{A} whose eigenvectors are linear independent can be decomposed into (see e.g. [2])

$$\mathbf{A} = \mathbf{V} \mathbf{D}_\lambda \mathbf{V}^{-1} \quad (8)$$

$$\mathbf{A} = \mathbf{V} \mathbf{D}_\lambda \mathbf{U} \quad (9)$$

where the matrix \mathbf{V} has the eigenvectors in its columns, and matrix \mathbf{D}_λ has the eigenvalues in its diagonal, and where

$$\mathbf{U} = \mathbf{V}^{-1} \quad (10)$$

The matrices in eq.(9) are given as

$$\mathbf{V} = \begin{bmatrix} \uparrow & \uparrow & \dots & \uparrow \\ \mathbf{v}_1 & \mathbf{v}_2 & \dots & \mathbf{v}_N \\ \downarrow & \downarrow & \dots & \downarrow \end{bmatrix}_{N \times N}, \quad \mathbf{D}_\lambda = \begin{bmatrix} \lambda_1 & 0 & \dots & 0 \\ 0 & \lambda_2 & \dots & 0 \\ \vdots & & \ddots & 0 \\ 0 & 0 & \dots & \lambda_N \end{bmatrix}_{N \times N}, \quad \mathbf{U} = \begin{bmatrix} \leftarrow & \mathbf{u}_1^T & \rightarrow \\ \leftarrow & \mathbf{u}_2^T & \rightarrow \\ \vdots & & \\ \leftarrow & \mathbf{u}_N^T & \rightarrow \end{bmatrix}_{N \times N} \quad (11)$$

From eq.(8), (9) and (11)

$$\mathbf{A}^k = \mathbf{V} \mathbf{D}_\lambda^k \mathbf{U} \quad (12)$$

$$= \sum_{i=1}^N \lambda_i^k \mathbf{v}_i \mathbf{u}_i^T \quad (13)$$

From eq.(7) and (13),

$$\mathbf{x}(k) = \sum_{i=1}^N \lambda_i^k \mathbf{v}_i \mathbf{u}_i^T \mathbf{x}(0) + \sum_{i=1}^N \sum_{j=1}^{k-1} \lambda_i^j \mathbf{v}_i \mathbf{u}_i^T \mathbf{b} + \mathbf{b} \quad (14)$$

From the denominator of eq.(2), we define the interference vector, $\mathbf{J}(k)$, as follows

$$\mathbf{J}(k) = (\mathbf{A} - \mathbf{D}_\mathbf{A})\mathbf{x}(k) + \mathbf{b} \quad (15)$$

where

$$\mathbf{D}_\mathbf{A} = \begin{bmatrix} a_{11} & 0 & \dots & 0 \\ 0 & a_{22} & \dots & 0 \\ \vdots & & \ddots & 0 \\ 0 & 0 & \dots & a_{NN} \end{bmatrix}_{N \times N} \quad (16)$$

Defining $\mu_{u,i} = \mathbf{u}_i^T \mathbf{x}(0)$ and $\mu_{b,i} = \mathbf{u}_i^T \mathbf{b}$, eq.(14) is rewritten as

$$\mathbf{x}(k) = \sum_{i=1}^N \lambda_i^k \mathbf{v}_i \mu_{u,i} + \sum_{i=1}^N \sum_{j=1}^{k-1} \lambda_i^j \mathbf{v}_i \mu_{b,i} + \mathbf{b} \quad (17)$$

Let's show the (single or multiple) eigenvalue whose norm is the greatest among the eigenvalues as λ_{max} . Then, in eq.(17), the term related to the λ_{max} , whose norm is greater than 1 by assumption, dominates the sum. This is because a relatively small increase in λ_j causes exponential increase as time step evolves, which is shown in the following: Let's define the following ratio

$$\kappa(k) = \frac{(\lambda_i)^k}{(\lambda_i + \Delta\lambda)^k} \quad (18)$$

where $\Delta\lambda$ represents the decrease (increase). The $\kappa(k)$ in eq.(18) is plotted in Figure 2 in [5] for some different $\Delta\lambda$ values. The Figure 2 in [5] implies that the term related to the λ_{max} dominate the sum in eq.(17). So, there exists a finite step number k_T such that

$$\mathbf{x}(k) = \lambda_{max}^k \mathbf{v}_m \mu_{u,m} + \sum_{j=1}^{k-1} \lambda_{max}^j \mathbf{v}_m \mu_{b,m}, \quad k \geq k_T \quad (19)$$

where \mathbf{v}_m is the eigenvalue corresponding to the λ_{max} . From eq.(7), (12) and (15)

$$\mathbf{J}(k) = \mathbf{V}\mathbf{D}_\lambda\mathbf{U}\mathbf{x}(k) - \mathbf{D}_\mathbf{A}\mathbf{x}(k) + \mathbf{b} \quad (20)$$

$$= \mathbf{V}\mathbf{D}_\lambda^{k+1}\mathbf{U}\mathbf{x}(0) + \sum_{j=1}^{k-1} \mathbf{V}\mathbf{D}_\lambda^{j+1}\mathbf{U}\mathbf{b} + \mathbf{V}\mathbf{D}_\lambda\mathbf{U}\mathbf{b} - \mathbf{D}_\mathbf{A}\mathbf{x}(k) + \mathbf{b} \quad (21)$$

Using eq.(14) and (15)

$$\mathbf{J}(k) = \sum_{i=1}^N \lambda_i^{k+1} \mathbf{v}_i \mathbf{u}_i^T \mathbf{x}(0) + \sum_{i=1}^N \sum_{j=1}^{k-1} \lambda_i^{j+1} \mathbf{v}_i \mathbf{u}_i^T \mathbf{b} + \mathbf{V}\mathbf{D}\mathbf{U}\mathbf{b} - \mathbf{D}_\mathbf{A}\mathbf{x}(k) + \mathbf{b} \quad (22)$$

As in the steps from eq.(17) to eq.(19), we see in eq.(22) that the term related to the maximum eigenvalue λ_{max} in magnitude, (which is greater than 1 by assumption), dominates the sum in (22). In other words, comparing eq.(17), (19) and (22), we obtain that

$$\mathbf{J}(k) = \lambda_{max}^{k+1} \mathbf{v}_m \mu_{u,m} + \sum_{j=1}^{k-1} \lambda_{max}^{j+1} \mathbf{v}_m \mu_{b,m} - \mathbf{D}_\mathbf{A}\mathbf{x}(k), \quad k \geq k_T \quad (23)$$

where $\mu_{u,m} = \mathbf{u}_m^T \mathbf{x}(0)$ and $\mu_{b,m} = \mathbf{u}_m^T \mathbf{b}$ as defined in eq.(19). Using eq.(19) in (23)

$$\mathbf{J}(k) = \{\lambda_{max} \mathbf{I} - \mathbf{D}_\mathbf{A}\} \mathbf{x}(k), \quad k \geq k_T \quad (24)$$

In eq.(23), we assume that $\mu_{u,m} \neq 0$, i.e., $\mathbf{x}(0)$ is not completely perpendicular to \mathbf{u}_m defined above. That's why we say "for "almost" any initial vector" in the proposition phrase. However, this is something easy to check in advance. If it is the case, then this can easily be overcome by introducing a small random number to $\mathbf{x}(0)$ so that it's not completely perpendicular to the \mathbf{u}_m . So, from eq.(24),

$$\frac{x_i(k)}{J_i(k)} = \frac{1}{\lambda_{max} - a_{ii}}, \quad k \geq k_T \quad (25)$$

We conclude from eq.(25) and from the "SIR" definition in eq.(2) that

$$\theta_i(k) = \frac{a_{ii}x_i(k)}{b_i + \sum_{j=1, j \neq i}^N a_{ij}x_j(k)} = \frac{a_{ii}}{\lambda_{max} - a_{ii}}, \quad k \geq k_T, \quad i = 1, \dots, N, \quad (26)$$

where λ_{max} is the (single or multiple) eigenvalue of \mathbf{A} with the maximum norm and k_T shows the finite time constant during which the λ_{max} become dominant in the sum of (17), which completes the proof.



Definition: State-specific ultimate SIR value: In proposition 1, we showed that the SIR in (2) for every state in the autonomous discrete-time linear networks in eq.(5) converges to a constant value as step number goes to infinity. We call this converged constant value as "ultimate SIR" and denote as θ_i^{ult} , $i = 1, 2, \dots, N$.

Does the result in the proposition above have any practical meanings? Our answer is yes. In [4] and [5], we present some (continuous-time and discrete-time) networks which are stabilized by the ultimate SIR as applied to the binary associative memory tasks as compared to the traditional Hopfield Neural Networks. So, similar to the networks in [5], the proposed autonomous network here is

$$\mathbf{x}(k+1) = \mathbf{A}\mathbf{x}(k)\delta(\theta(k) - \theta^{ult}) \quad (27)$$

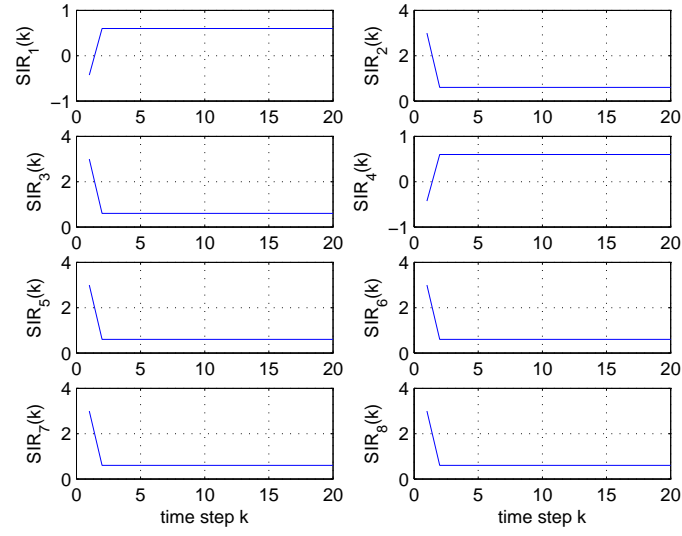
$$\mathbf{y}(k) = \text{sign}(\mathbf{x}(k)) \quad (28)$$

where \mathbf{I} is the identity matrix, \mathbf{A} is the system matrix, $\theta^{ult} = [\theta_1^{ult}\theta_2^{ult}\dots\theta_N^{ult}]^T$, and $\theta(k) = [\theta_1(k)\theta_2(k)\dots\theta_N(k)]^T$ is the SIR vector at step k whose elements are given by eq.(2), the function $\delta(\cdot)$ gives 0 if and only if its argument vector is equal to zero vector, and gives 1 otherwise, and $\mathbf{y}(k)$ is the output of the network.

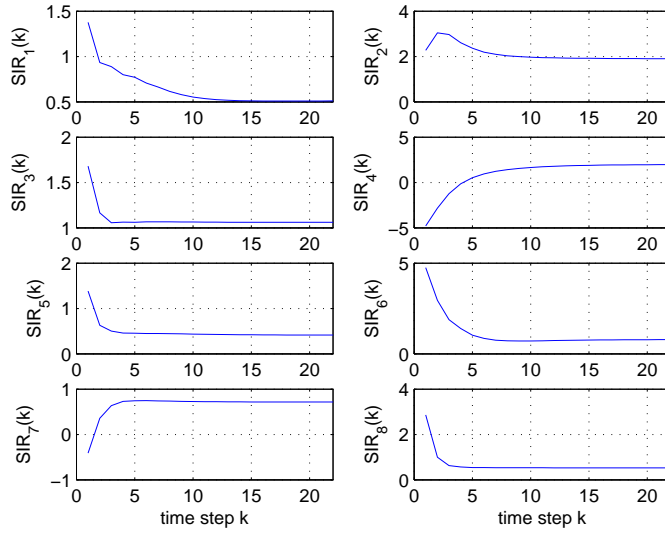
The proof of the proposition above shows that in the linear network of (5), the defined SIR in eq.(2) for state i converges to the constant SIR value in eq. (6), for any initial condition $\mathbf{x}(0)$ within a finite step number k_T . It's seen that the linear networks of eq.(5) is nothing but the underlying network of the proposed network without the $\delta(\cdot)$ function. Since the "SIR" in eq.(2) exponentially approaches to the constant Ultimate "SIR" in eq.(6), the delta function will stop the exponential increase once $\theta(k) = \theta^{ult}$, at which the system output reach its steady state response. So, the the presented network is stable. Furthermore, we note that the proposed networks in [4] and [5] which have been applied to the binary associative memory systems are special cases of the proposed network in eq.(27) and (28).

III. SIMULATION RESULTS

In this section, we borrow the simple simulation examples in [5] with 8 and 16 neurons. We plot some samples of the evolution of the SIRs with respect to step number in the system of



(a)



(b)

Fig. 1. A sample evolution of $SIR(k)$ with respect to time step for 8-dimensional system with (a) the weight matrix in [5] when the initial vector is 2-Hamming distance away from desired vector, and (b) random weight matrix.

[5] whose weight matrix is presented in [5] as well as in the system of eq.(5) whose weight matrix is chosen randomly for illustration purpose. Figure 1.a shows a sample of SIR evolution in the 8-neuron case in the example 1 of [5], for initial condition $[-111 - 1 - 1 - 1 - 1 - 1]$

while the desired vector is $[1111 - 1 - 1 - 1 - 1]$ (Hamming distance is 2). Figure 1.b shows a sample evolution of SIR in case of the system in eq.(5) with a random weight matrix. Some sample evolution plots for the 16-neuron case are shown in Figure 2 for the system of eq.(5) with random weight matrix. The figures 1 and 2 show that the SIRs converge to their ultimate values and confirm the results in previous section.

IV. CONCLUSIONS

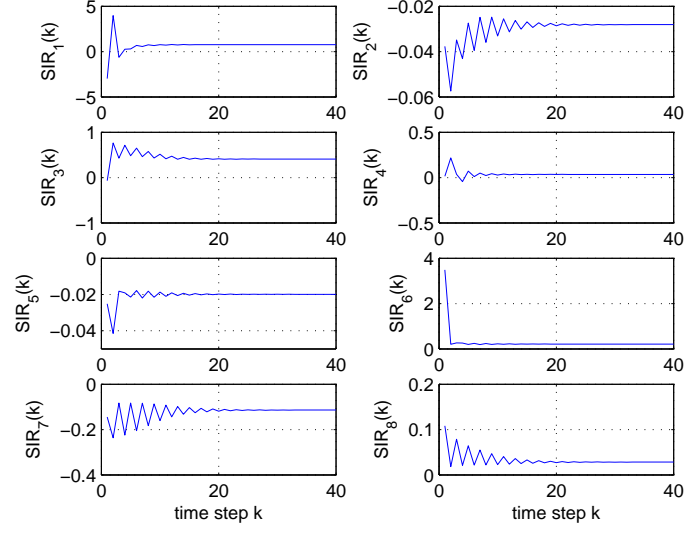
In this letter, we improve the SIR results in [5] by relaxing the symmetry assumption and taking also the noise term into account. We examine the following discrete-time autonomous linear system: $\mathbf{x}(k+1) = \mathbf{A}\mathbf{x}(k) + \mathbf{b}$, where \mathbf{A} is any real square matrix with linearly independent eigenvectors whose largest eigenvalue is real and its norm is larger than 1, and vector \mathbf{b} is constant. Using the same “SIR” concept as in [4] and [5], we show that the ultimate “SIR” is equal to $\frac{a_{ii}}{\lambda_{max} - a_{ii}}$, $i = 1, 2, \dots, N$, where N is the number of states, λ_{max} is the eigenvalue with maximum norm and a_{ii} is the diagonal elements of matrix \mathbf{A} .

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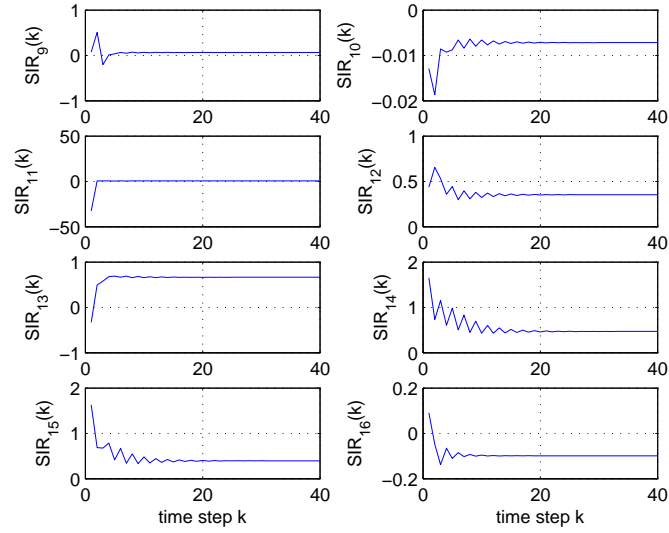
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REFERENCES

- [1] T.S. Rappaport, *Wireless Communications: Principles and Practice*, Prentice-Hall, New York, 1996.
- [2] O. Bretscher, *Linear Algebra with Applications*, Prentice Hall, 2005.
- [3] Z. Uykan, “From Sigmoid Power Control Algorithm to Hopfield-like Neural Networks: “SIR” (“Signal”-to-“Interference”-Ratio)- Balancing Sigmoid-Based Networks- Part II: Discrete Time”, *submitted to IEEE Trans. Neural Networks*, December, 2008.
- [4] Z. Uykan, “Ultimate “SIR” (“Signal”-to-“Interference”-Ratio) in Continuous-Time Autonomous Linear Networks with Symmetric Weight Matrices, and Its Use to “Stabilize” the Network as Applied to Binary Associative Memory Systems”, *submitted to IEEE Transactions on Circuits and Systems I*, February, 2009.
- [5] Z. Uykan, “ Analysis of “SIR” (“Signal”-to-“Interference”-Ratio) in Discrete-Time Autonomous Linear Networks with Symmetric Weight Matrices”, *submitted to IEEE Transactions on Circuits and Systems I*, February, 2009.



(a)



(b)

Fig. 2. A sample evolution of $SIR(k)$ with respect to time step for 16-dimensional system with random weight matrix, (a) SIRs from 1 to 8, (b) SIRs from 9 to 16.

